

# DANSETTE RT222

meter. The receiver was switched to M.W. and the volume control was set in the minimum output position. All voltages are negative with respect to chassis.

Transistor Table

Transistor	Emitter (V)	Base (V)	Collector (V)
TR1 OC44 ..	1.1	1.0	7.0
TR2 OC45 ..	0.5	0.8	7.1
TR3 OC45 ..	0.9	1.1	7.1
TR4 OC81D ..	1.1	1.2	8.6
TR5 OC81 ..	—	0.2	8.9
TR6 OC81 ..	—	0.2	8.9

Battery consumption was 13.5mA with no signal input; 25mA with normal signal. **Alternative Transistor Types.**—Some receivers are fitted with a range of G.E.C. transistors in place of Mullard as follows: TR1 GET874, TR2 GET873, TR3 GET873, TR4, TR5 and TR6 GET114. In these receivers R16 becomes 6.8kΩ. In early production receivers where Mullard transistors are fitted, TR4 may be OC78D, TR5 and TR6 may be OC78.

### CIRCUIT DESCRIPTION

Signal input is by way of the internal aerial L2/L4 (L.W.) and L3/L5 (M.W.) or from an external aerial via coupling winding L1. Switches S1, S2, S4 and S5 select the appropriate coil windings; S3 short-circuits the L.W. aerial coil on M.W. Tuning is by C1 and C2 on both wavebands with C3 added in parallel on L.W. R.F. signals are fed directly to the base of TR1 which operates as a self-oscillating mixer. Base bias is derived from the potential divider R1, R2, L8 with C8 and C9 (M.W.) comprise the

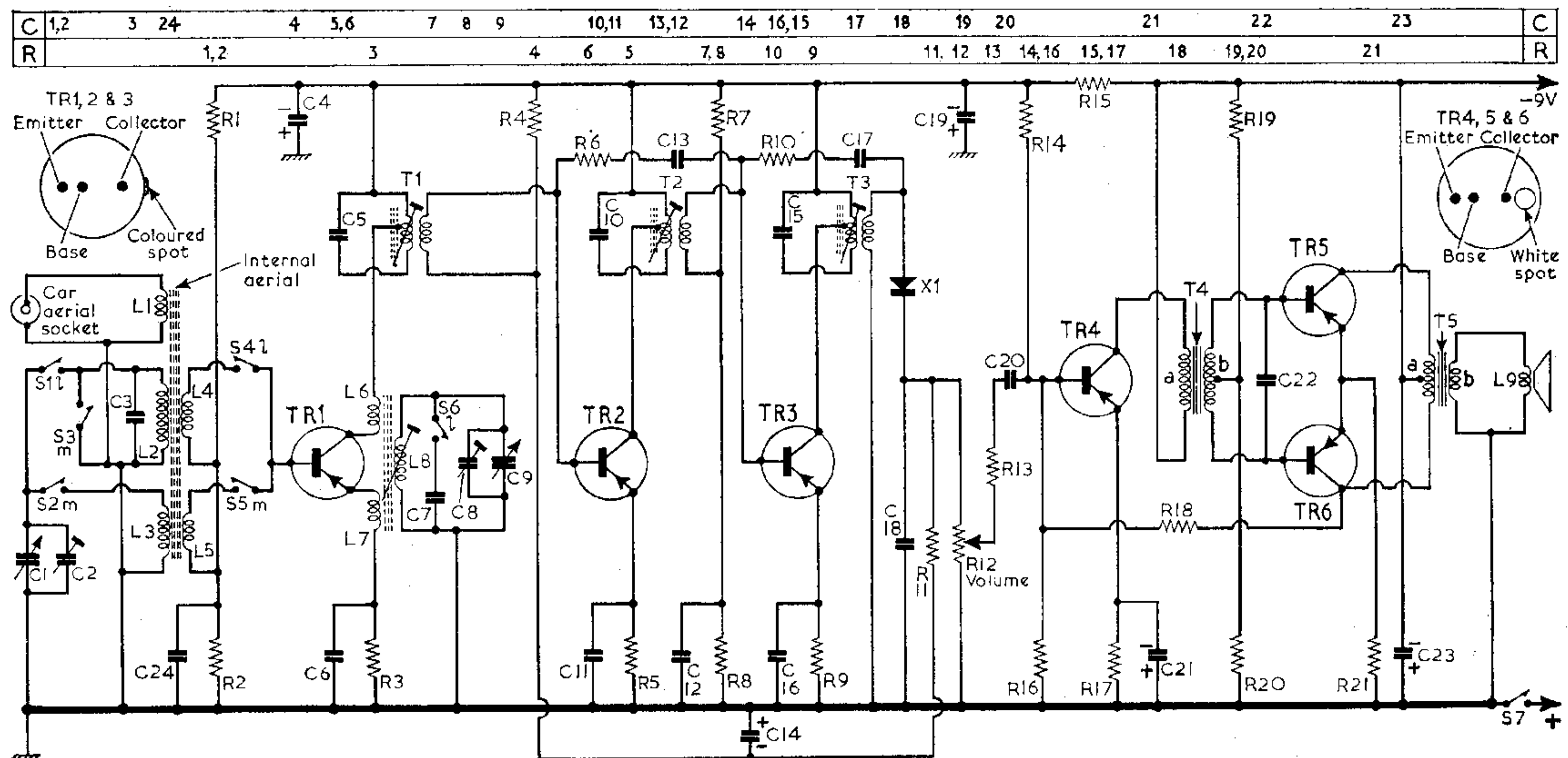
(Continued overleaf col. 1)

**DANSETTE RT222** is a six-transistor portable radio receiver designed for Medium and Long wave reception and housed in a plastics fabric covered, glass-fibre case. It is fitted with an internal ferrite-rod aerial and is provided with a socket for the connection of a car aerial. The chassis comprises two printed-circuit panels (RF and audio) which together with the loudspeaker are secured to a metal frame. The whole is removable as a single unit from the case for servicing. Operation is from a single 9V battery. Release date and original price: March 1961, £12 0s 7d. Purchase tax extra.

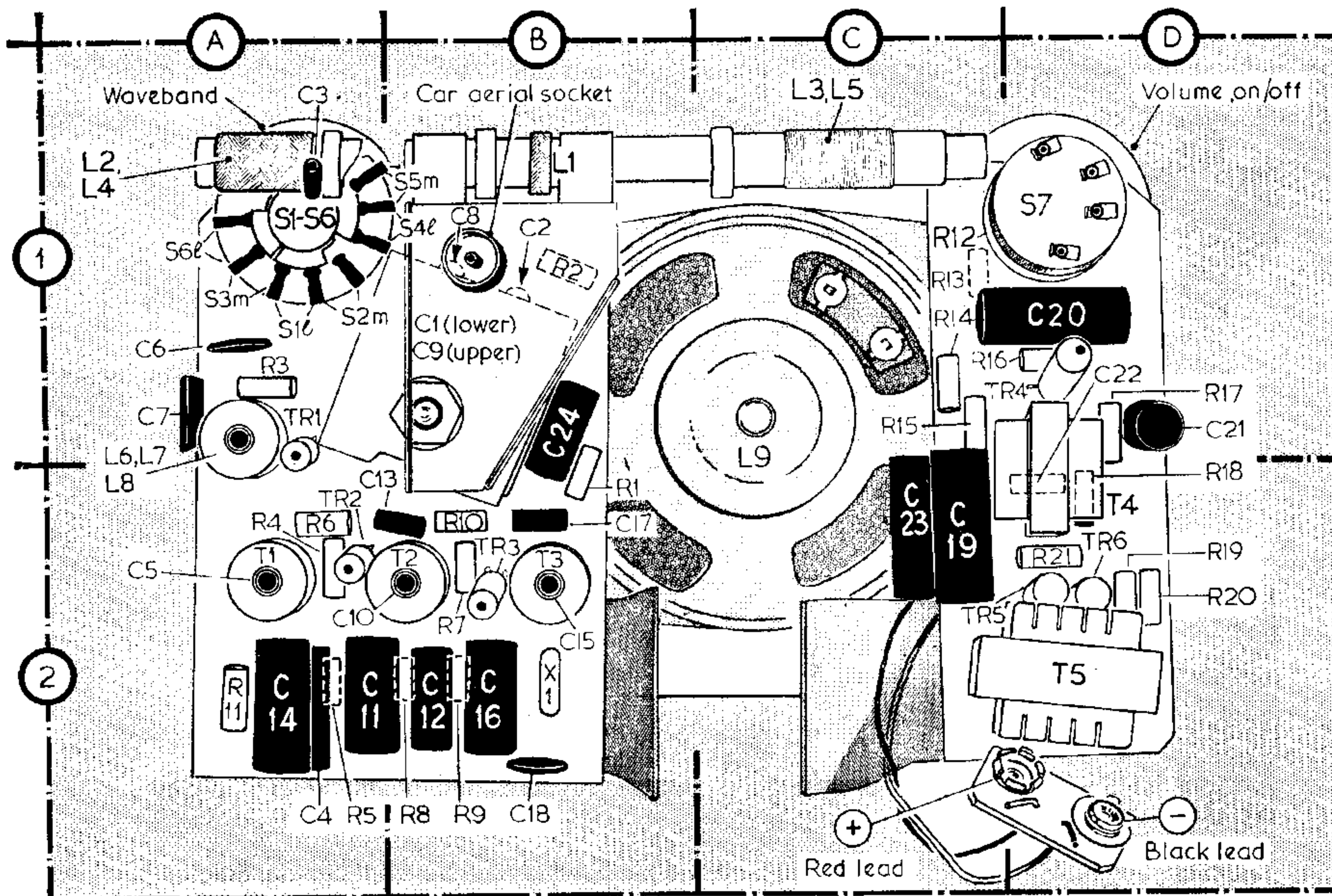
### TRANSISTOR ANALYSIS

Transistor voltages given in the table in col. 2 were measured on our specimen receiver using an Avo Electronic Test-

Resistors			Capacitors			Coils*			Transformers*			Miscellaneous		
R1	56kΩ	B2	C1	157pF	B1	L1	1.25	B1	T1	—	A2	X1	OA70 or GEX34	B2
R2	10kΩ	B1	C2	20pF	B1	L2	7.5	A1	T2	—	A2	S1-S6	—	A1
R3	3.9kΩ	A1	C3	18-56pF	A1	L3	1.5	C1	T3	—	B2	S7	—	D1
R4	68kΩ	A2				L4	—	A1	T4	{ a 115.0 } { b 86.0 }	D2			
R5	680Ω	A2							T5	{ a 4.0 } { b — }	D2			
R6	1.2kΩ	A2												
R7	22kΩ	B2												
R8	4.7kΩ	B2												
R9	1kΩ	B2												
R10	3.9kΩ	B2												
R11	8.2kΩ	A2												
R12	5kΩ	C1												
R13	2.7kΩ	C1												
R14	39kΩ	C1												
R15	680Ω	C1												
R16	12kΩ†	D1												
R17	560Ω	D1												
R18	1MΩ	D2												
R19	4.7kΩ	D2												
R20	100Ω	D2												
R21	4.7Ω	D2												



Circuit diagram of Dansette RT222. In some receivers a 560Ω or 680Ω resistor may be wired across T5 primary



The complete receiver removed from its case, observed from the rear. Components not directly visible from this angle are trimmers C2 and C8 which are mounted on the tuning gang behind the aerial socket plate (location reference B1), also R13 and C22 which are mounted on the reverse side of the audio panel (D1)

**Circuit Description—continued**

oscillator tuned circuit generating the heterodyne signal which is injected via L7 into the emitter circuit. Regenerative feedback coupling from collector to emitter is provided by L6 and L7. C7 is brought into circuit by S6 on L.W. to tune L8 to the lower frequency waveband.

Intermediate frequency selected by T1 is at 470 kc/s and is coupled via the secondary of the single tuned transformer to the base of TR2. TR2 and TR3 operate as a two-stage I.F. amplifier in conjunction with associated coupling transformers T2 and T3.

The primary windings of T1, T2 and T3 are suitably tapped to match the low impedance collector output of the transistors. R6, C13 and R10, C17 are feedback circuits which neutralize the internal capacitances of TR2 and TR3. Amplified I.F. output from TR3 is fed to the detector diode X1 and rectified audio signals are developed across the combined load resistor and volume control R12. The D.C. component present across R12 is tapped off and after filtering by R11 and C14 is returned to the base of TR2 as A.G.C. bias.

From the slider of R12 the audio signal is coupled via C20 to the base of the driver TR4. Connected in the collector circuit of TR4 the driver transformer T4 has a centre-tapped secondary which feeds the bases of TR5 and TR6 in anti-phase. These two transistors form a class B push-pull output stage which drives the loudspeaker L9.

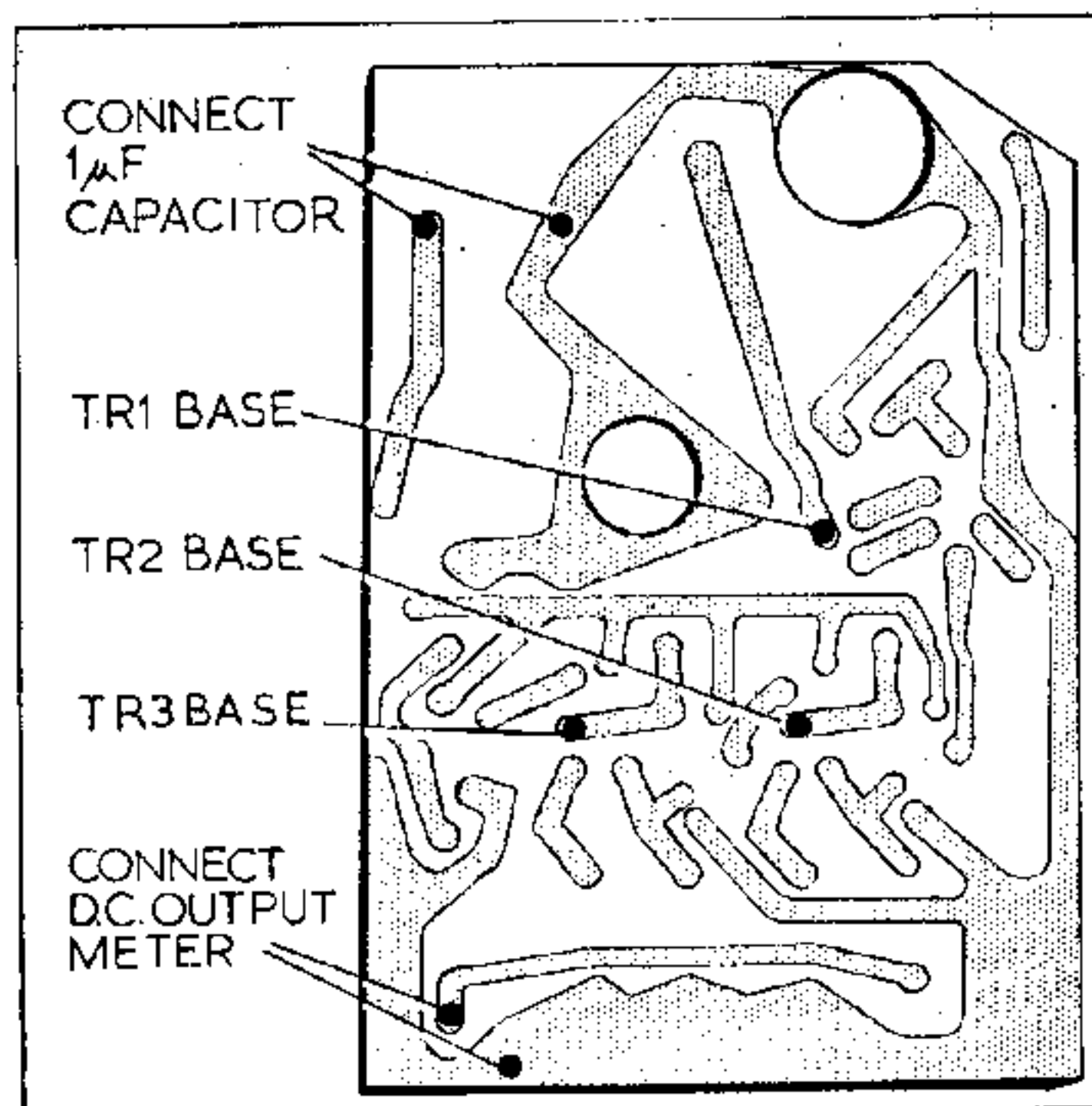
**CIRCUIT ALIGNMENT**

**Equipment Required.**—An A.M. signal generator; a high resistance 0-2.5V D.C. voltmeter; an R.F. coupling loop constructed by winding 3 turns of insulated wire to a diameter of 10ins. with a 430Ω resistor connected in series; a 0.5μF capacitor and an 820 ohm resistor wired in series for use as a generator terminating network; a 1μF capacitor and a bladed type trimming tool for the I.F. transformer cores.

The identity of the equipment connecting points referred to in the instructions which

follow, are given in the illustration of the foil side of the receiver panel below.

- 1.—Connect the D.C. voltmeter as an audio output meter across the volume control R12. Switch receiver to M.W. and turn the volume control to minimum output.
- 2.—Connect the signal generator via the 0.5μF capacitor and 820 ohm resistor network to the base of TR3. Feed in a 470 kc/s signal and adjust T3 (location reference B2) for maximum output on the D.C. meter.
- 3.—Transfer the signal generator to the base of TR2 and adjust T2 (A2) for maximum output.
- 4.—Shunt C24 (B1) with the 1μF capacitor. Transfer the signal generator to the base of TR1 and adjust T1 (A2) for maximum output.
- 5.—Rotate the tuning gang to the maximum capacitance position (fully meshed). Feed in a 540 kc/s signal and adjust L8 (A1) for maximum output.
- 6.—Rotate the tuning gang to the minimum capacitance position. Feed in a 1,640 kc/s signal and adjust C8 (B1) for maximum output.
- 7.—Repeat operations 5 and 6 until no further improvement can be obtained. Remove the 1μF capacitor from across C24.
- 8.—Connect the signal generator output leads across the R.F. coupling loop and place the loop about 2 feet distant from the ferrite rod. Tune receiver to the 500m mark on the scale. Feed in a 600 kc/s signal and adjust L3 (C1) for maximum output.
- 9.—Tune receiver to the 250m mark on the scale. Feed in a 1,200 kc/s signal and adjust C2 (B1) for maximum output.
- 10.—Repeat operations 8 and 9 until no further improvement can be obtained.
- 11.—Switch receiver to L.W. and tune to 1,400m on scale. Feed in a 214 kc/s signal and adjust L2 (A1) for maximum output.



Foil side of R.F. panel showing connections required for alignment purposes

**GENERAL NOTES**

**Dismantling.**—To remove the chassis from the cabinet first undo two coin-slotted screws and remove the back cover.

Remove and disconnect the battery. Pull off the tuning control knob. Remove two hexagonal pillars which secure the chassis.

Using the loudspeaker magnet as a grip, lift out the chassis in a down and outwards movement to clear the case with the control knobs.

**Switches.**—S1-S6 are the waveband switches which are housed in a two-way rotary unit shown in location reference A1, where individual switch sections can be identified. In the switch drawing and on the circuit diagram, suffix letter m means closed on M.W. and suffix letter l means closed on L.W. S7 is the battery on/off switch and is ganged with the volume control R12.

**Battery.**—9V Vidor T6007, Ever-Ready PP7, or equivalent.

**Modifications.**—In earlier receivers R20 was 82Ω not 100Ω.